

Simulation of a Chain of Clocks



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Outline of Presentation

- ▶ Some Background
- ▶ Chain of clocks
- ▶ Mathematical Model for analysis
- ▶ Frequency model for the oscillator
- ▶ Analysis of clock error (Phy layer) for chain of 5 T-BCs
- ▶ Spectral models appropriate for SEC and eSEC network limit MTIE masks & frequency deviation in endpoint clock when output of chain is at network limit
- ▶ Concluding Remarks

Some Background

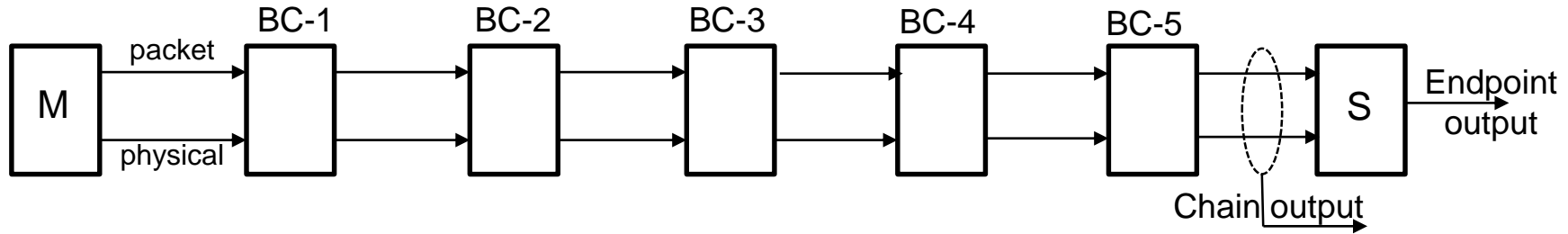
► Motivations for this study:

- Analyze the timing behavior of a chain of clocks and provide a rapid (not time-consuming) approach to estimate timing behavior
- The focus of this study is the physical layer chain with emphasis on the end-point (application) clock

► The underlying premise:

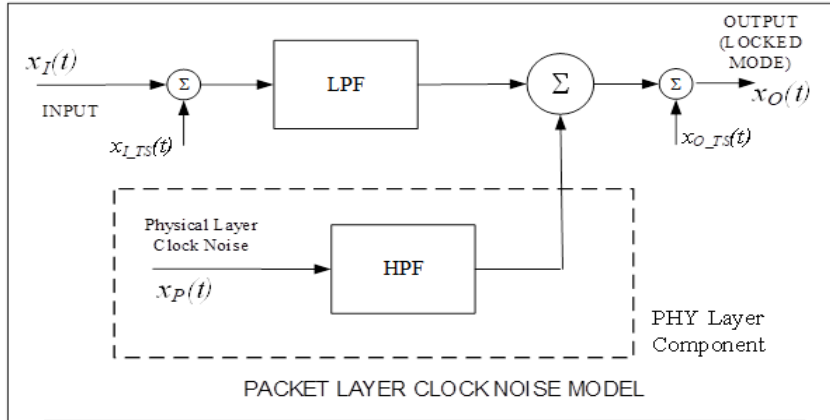
- Analytical approach to evaluation of clock performance in a chain of clocks using a frequency domain approach to estimate standard deviation
- Other Metrics (MTIE, $\max|TE|$) involving “peak” and “maximum” estimated as G times standard deviation (G is about 4.0 to 8.0) derived from spectrum
- All noise sources are uncorrelated so that they add in power (variance)
- **Frequency offsets, constant Time Error (cTE), asymmetry effects, deterministic signals, temperature effects, and time-varying phenomena are not included here**

Chain of clocks (Fronthaul, C-RAN)

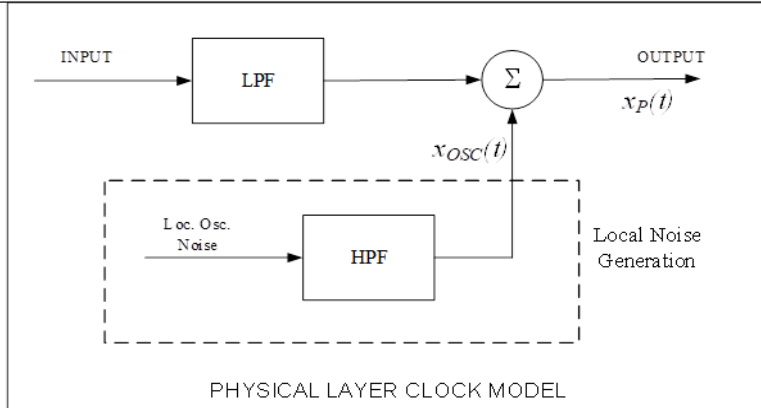


- ▶ Timing chain for Fronthaul/C-RAN likely to be less than 5 boundary clocks between Master (Edge GrandMaster or BBU) and Slave (Small Cell / RRH)
- ▶ The output of the chain must meet the requirements (e.g. TDEV, MTIE) of the physical layer exemplified by the SEC or eSEC network limit
- ▶ The end-point (application) clock must meet the application requirements even when the chain output is at the network limit
- ▶ In this presentation we consider the physical layer (“SyncE”); the time layer (PTP) is considered separately

Mathematical model for analysis



- Time layer clock noise model
- “Input” of clock n is output of clock $(n-1)$ plus time-stamping noise of clock n
- Time-stamps T2 and T4 have quantization error (time-stamp granularity).



Focus of this Presentation

- Physical layer clock noise model
- Input of clock n is output of clock $(n-1)$
- Physical layer clock output of clock n also goes to time-layer of clock n
- Local noise generation modeled as either based on oscillator **or** based on noise-generation TDEV mask

Mathematical formulation

Assumed sampling rate $f_0 = 16\text{Hz}$; sampling interval $\tau_0 = 0.0625\text{s}$

$$\sigma_x^2(\tau = n \cdot \tau_0) = G(n) \cdot \int_0^{0.5 \cdot f_0} S_x(f) \cdot \frac{(\sin(n\pi f \tau_0))^6}{(\sin(\pi f \tau_0))^2} df \quad \text{TVAR from spectrum}$$

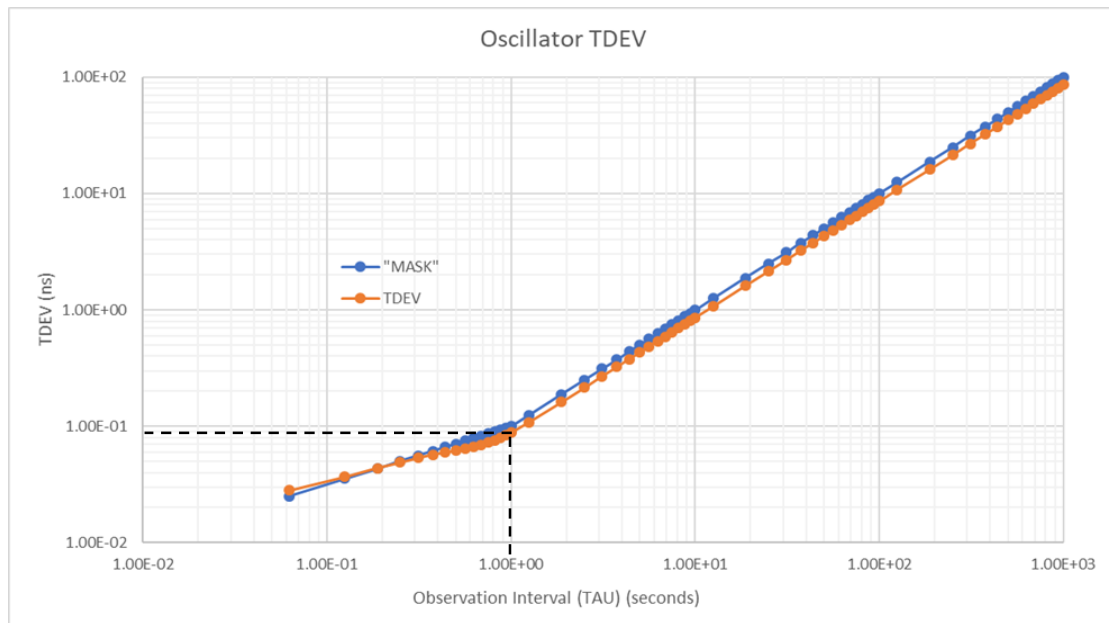
$$\sigma_{\xi}^2(n) \approx \int_0^{f_0} S_x(f) \cdot 4 \cdot (\sin(n\pi f \tau_0))^2 df$$

MTIE from spectrum
{ maximum = 7 · sigma }

$$M_x(\tau) \leq 7 \cdot \sigma_{\xi}(n)$$

- ▶ The calculation of frequency offset as a function of observation interval is a subset of the MTIE calculation
- ▶ MTIE calculation includes addressing the non-decreasing nature of MTIE

Oscillator Model

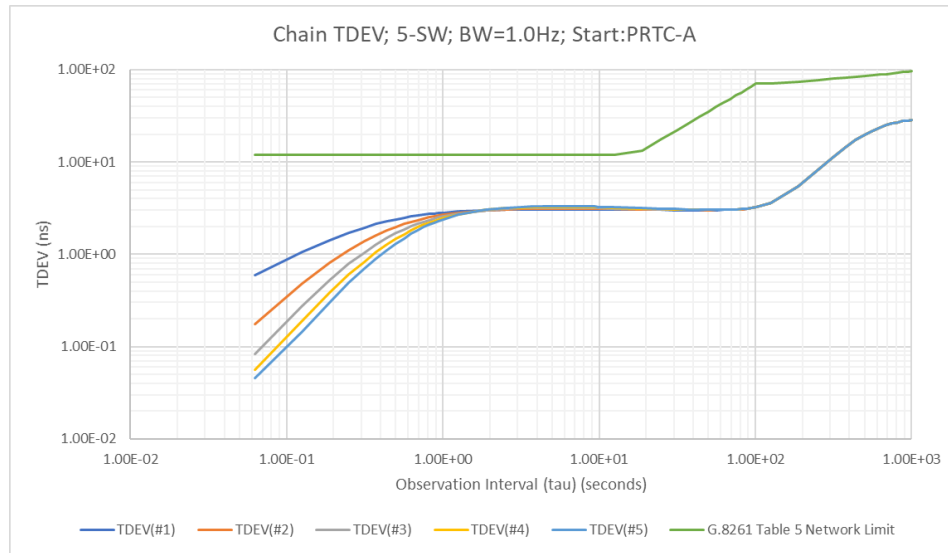


- Mixture of white FM (low tau) and flicker FM (larger tau)
- “knee” is the cross-over point, assumed to be 1s here
- Oscillator quality represented by TDEV(τ = knee)
- “Mask” = straight line view
- TDEV for very low observation intervals (\ll knee) represents phase-noise (not considered here)
- Oscillator spectrum constructed as low-pass-filtered flicker FM and high-pass filtered white FM and resulting TDEV shown

Note: white FM TDEV follows $\tau^{0.5}$; flicker FM TDEV follows $\tau^{1.0}$; range of τ considered is small enough to ignore higher order noise types.

Assumption: TDEV(@1s) = 0.1ns

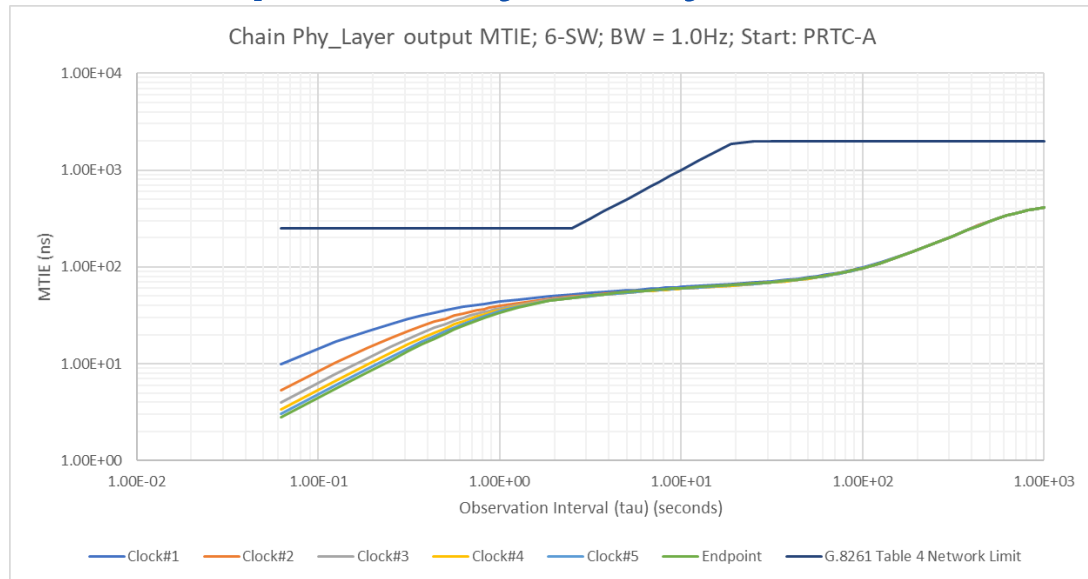
Chain output Phy_Layer - TDEV



Note: Principal source of clock noise is the starting point (PRTC-A)

- ▶ Expected TDEV of clock chain (5 T-BC) (phy_layer) using oscillator with assumed performance ($\text{TDEV}(@1\text{s}) = 0.1\text{ns}$) for clock bandwidth of 1.0Hz
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 5)
- ▶ Cases with other bandwidths provided in back-up

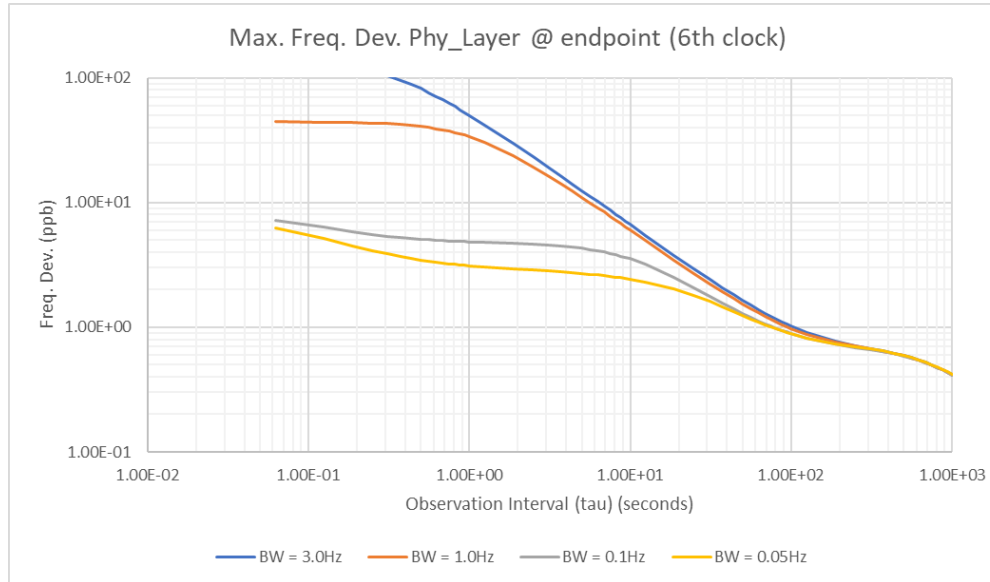
Chain output Phy_Layer - MTIE



Note: This simulation used the same bandwidth for all clocks – chain as well as endpoint.

- ▶ Expected MTIE of clock chain (phy_layer) (including endpoint) using oscillator with assumed performance ($TDEV(@1s) = 0.1ns$) for bandwidth of 1.0Hz
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 4)
- ▶ Cases with other bandwidths provided in back-up

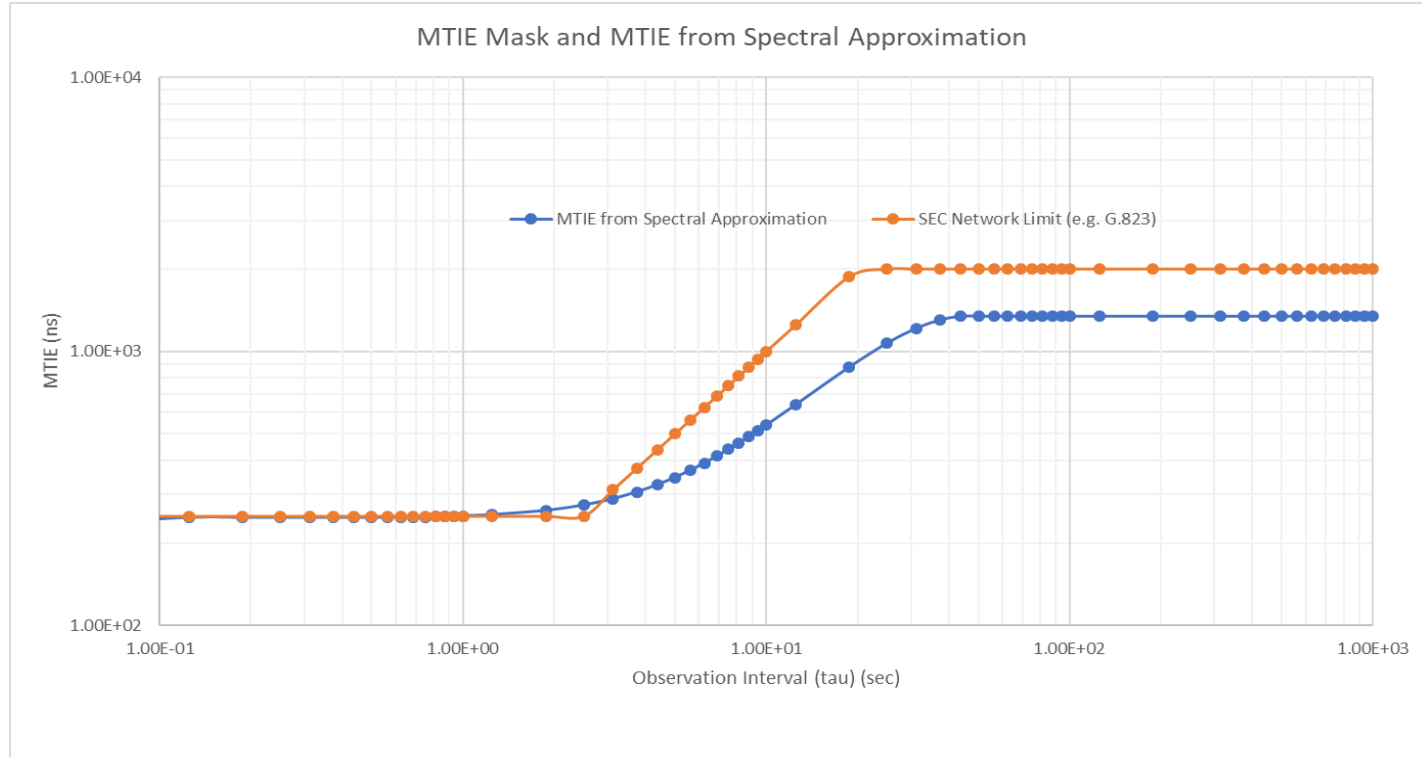
Max. Freq. Deviation (Phy_Layer) - Endpoint



Note: This simulation used the same bandwidth for all clocks – chain as well as endpoint.

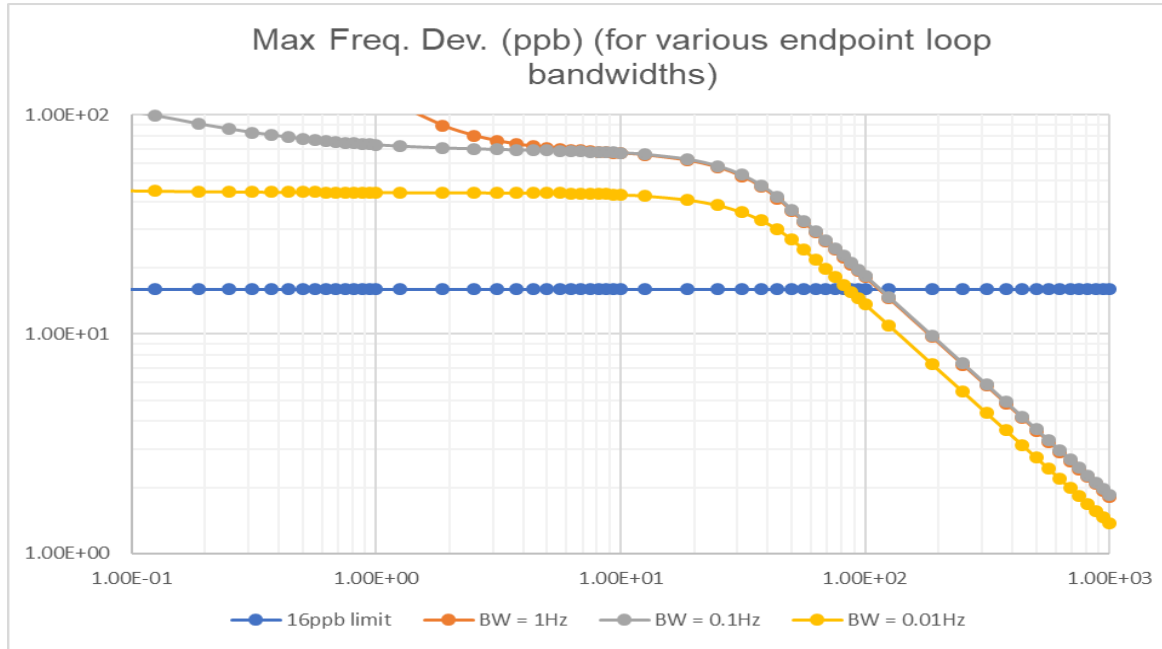
- ▶ Expected maximum frequency deviation (phy_layer) as a function of observation interval for various bandwidths (using oscillator with TDEV(@1s) = 0.1ns)
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ **Common requirement is “better than 16ppb”**
- ▶ Phy_layer clock often used to generate RF carrier in end-point (e.g. RRH)

Spectral Model Matching SEC MTIE Network Limit



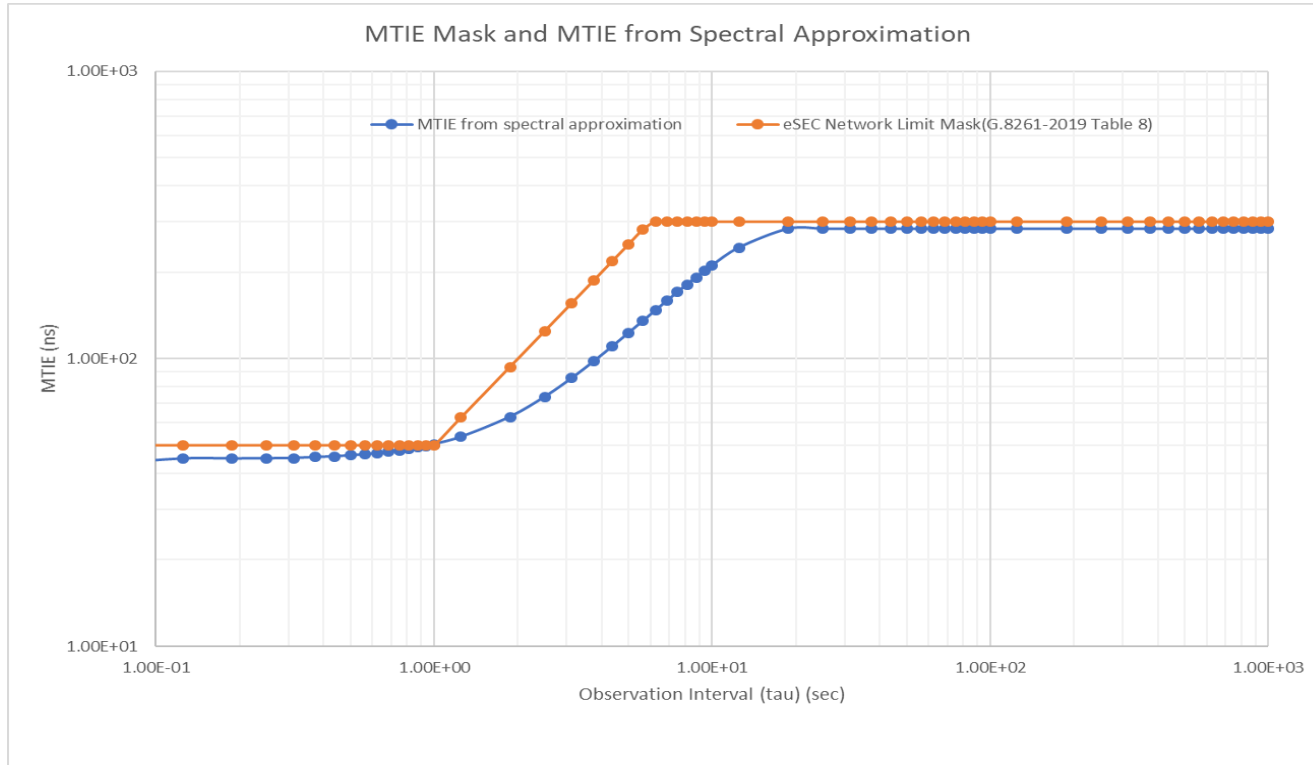
- Not possible to find a suitable $S(f)$ that yields a perfect MTIE match

Max. Freq. Deviation (Phy_Layer) - Endpoint



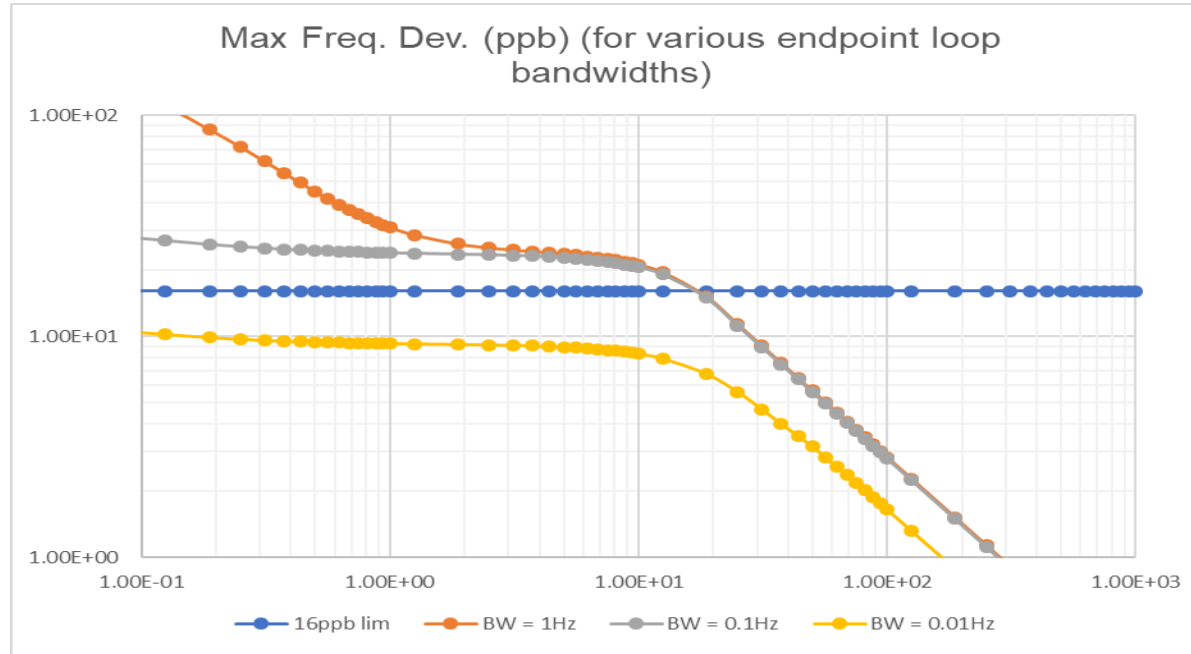
- ▶ Expected maximum frequency deviation (phy_layer) as a function of observation interval for various bandwidths (using oscillator with $TDEV(@1s) = 0.1ns$)
- ▶ Input to Endpoint set to regular SEC Mask (approximation)

Spectral Model Matching eSEC MTIE Network Limit



- Not possible to find a suitable $S(f)$ that yields a perfect MTIE match

Max. Freq. Deviation (Phy_Layer) - Endpoint



- ▶ Expected maximum frequency deviation (phy_layer) as a function of observation interval for various bandwidths (using oscillator with $TDEV(@1s) = 0.1ns$)
- ▶ Input to Endpoint set to enhanced SEC Mask (approximation)

Concluding Remarks

- ▶ The results described here used the frequency-domain method.
 - The frequency-domain method is computationally rapid and can provide results quickly, allowing analysis of several “what if?” scenarios in a timely fashion but cannot handle time-varying phenomenon
- ▶ A suitable model for an oscillator (random noise viewpoint):
 - Preferably involves 2 parameters:
 - “knee” : observation interval corresponding to cross-over between white FM and flicker FM
 - TDEV(@ τ = “knee”)
 - A single parameter, namely TDEV(τ = “knee” = 1s), may be adequate if actual knee > 1s.
- ▶ The method permits an analysis of the trade-offs between bandwidth and performance metrics
 - Example: to meet the maximum frequency deviation of 16ppb at the endpoint, for a chain starting with a G.811 quality clock, a bandwidth of the order of 0.05Hz is recommended
 - Example: when the input to the endpoint clock is at the regular SEC network limit, the 16ppb limit is achievable for observation intervals greater than 100s for various bandwidths.
 - Example: when the input is at the enhanced SEC network limit, the 16ppb limit is met for all observation intervals with a bandwidth of 0.01Hz.



Questions?
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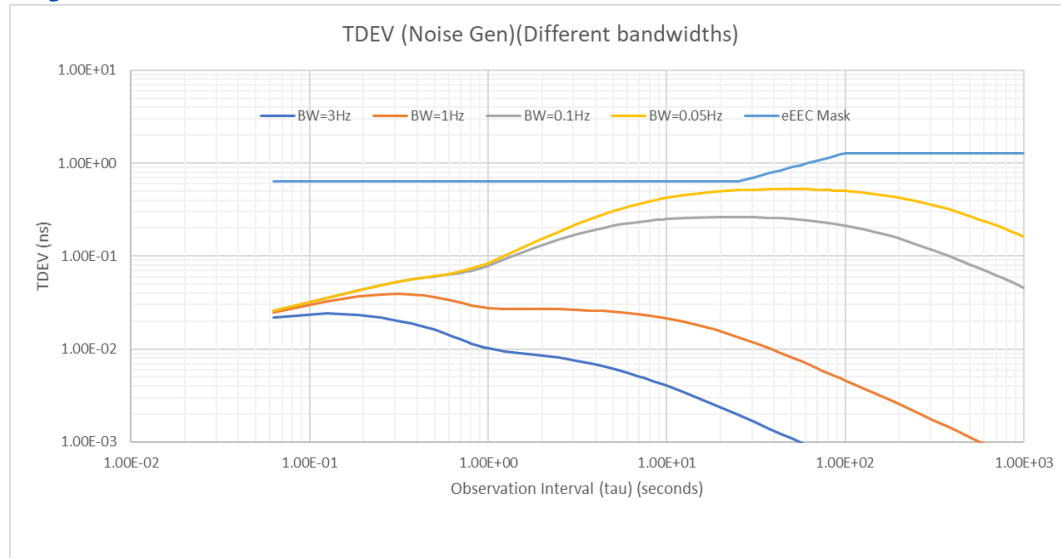
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Backup Slides

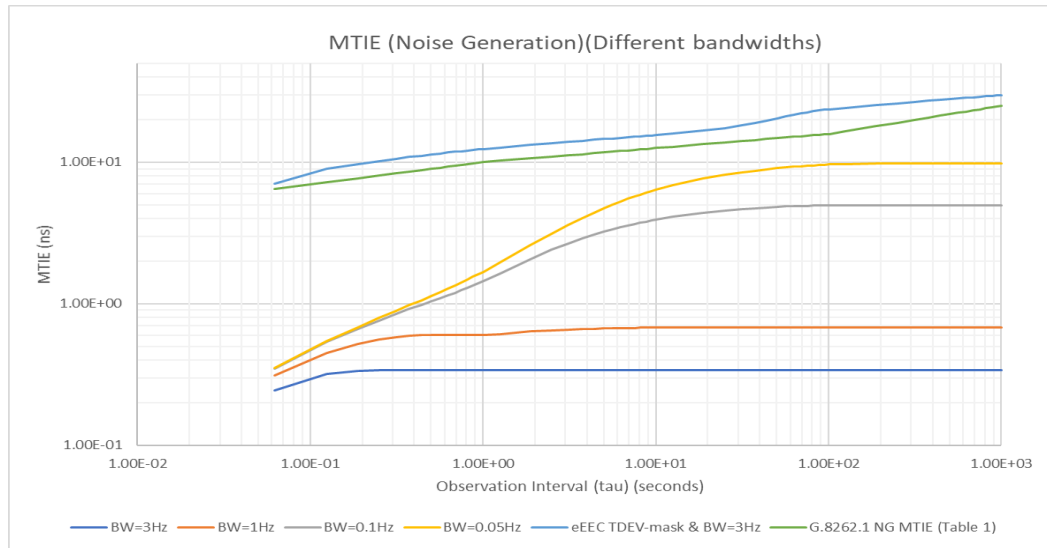
Additional Simulation Results

Phy_Layer Noise Generation - TDEV



- ▶ Expected noise generation of phy_layer clock using oscillator with assumed performance for various bandwidths
- ▶ Enhanced EEC bandwidth is specified as between 1Hz and 3Hz
- ▶ With this oscillator even bandwidth as low as 0.05Hz is compliant

Phy_Layer Noise Generation - MTIE



Note: Even if the clock noise generation satisfies the required TDEV (mask) it could violate the required MTIE mask, even with the widest bandwidth (3Hz). This is a known effect.

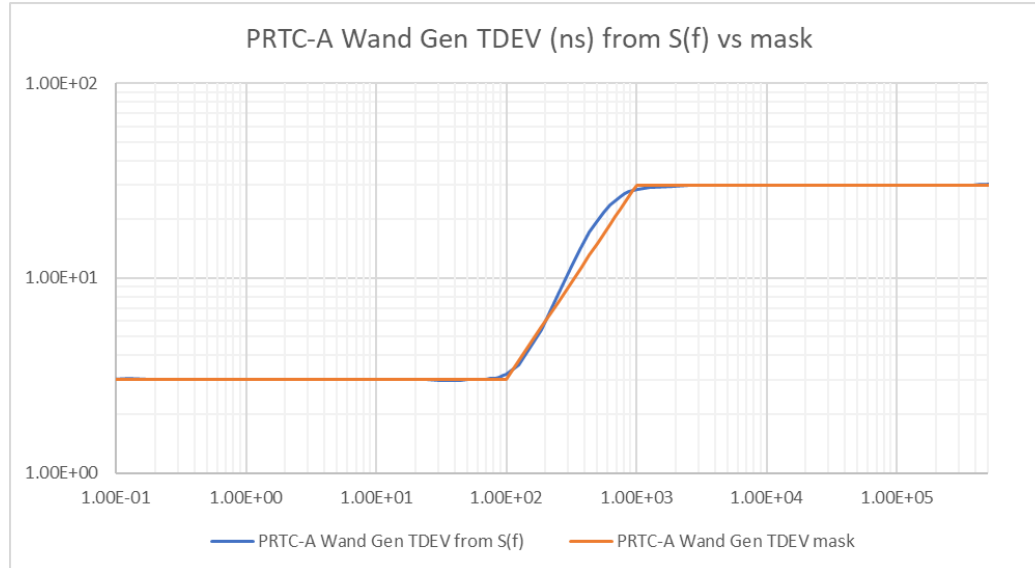
Note: MTIE calculation here is conservative.

- ▶ Expected noise generation of phy_layer clock using oscillator with assumed performance for various bandwidths
- ▶ Enhanced EEC bandwidth is specified as between 1Hz and 3Hz
- ▶ With this oscillator even bandwidth as low as 0.05Hz is compliant

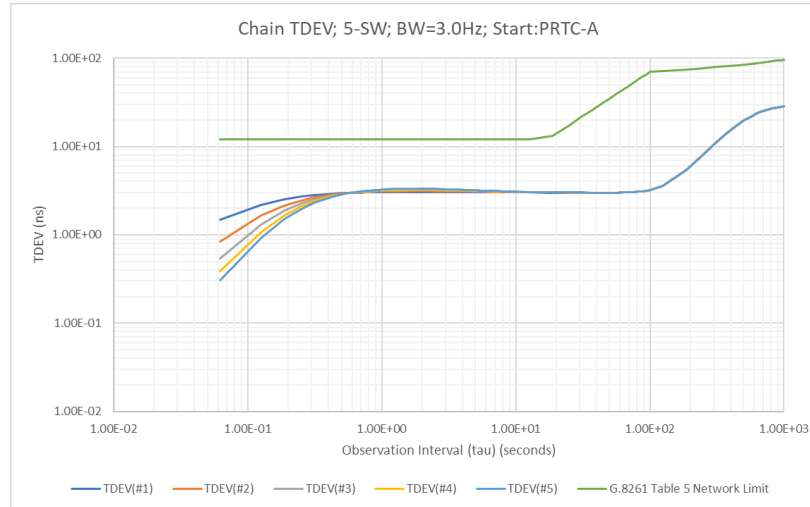
PRTC-A Wander Generation TDEV Mask

$$S_x(f) = \left(\frac{G}{f} \cdot (a_1 \cdot H_1(f) + a_2 \cdot H_2(f)) \right) \cdot C(f)$$

- $H_1(f)$ is a 2nd-order high-pass filter with 3-dB frequency of ~0.002Hz and $a_1 = 3.0^2$;
- $H_2(f)$ is a 3rd-order low-pass filter with 3-dB frequency of ~0.0007Hz and $a_2 = 30.0^2$;
- $G \sim 0.44$.

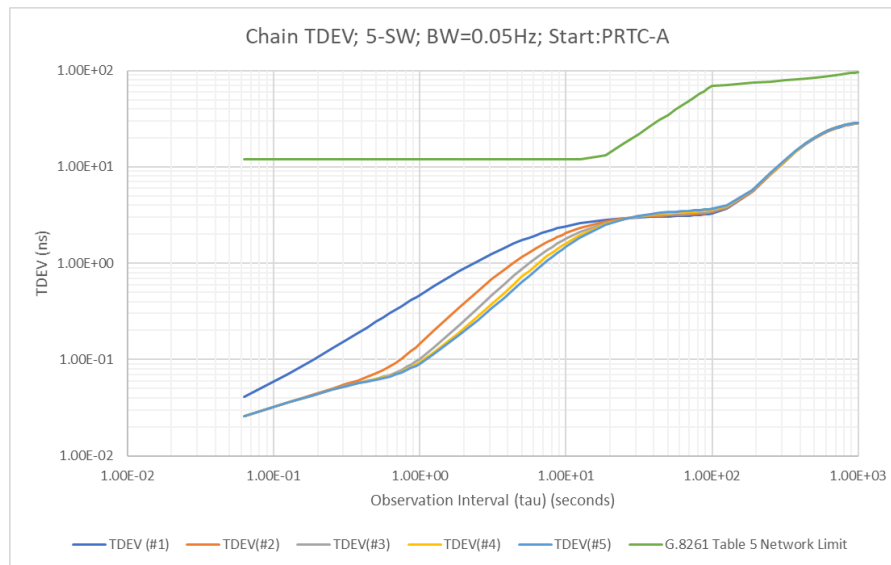


Chain output Phy_Layer - TDEV



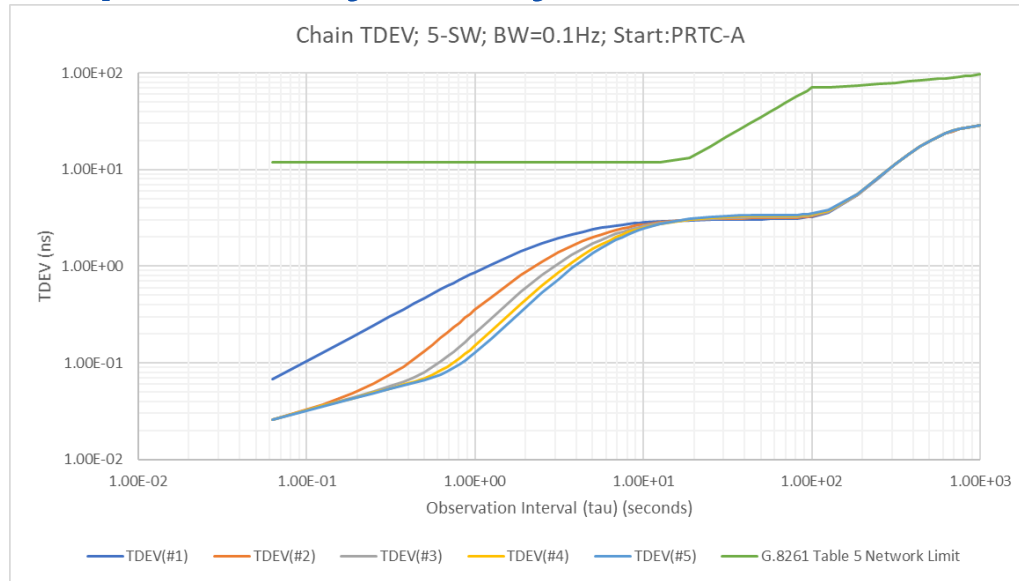
- ▶ Expected TDEV of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 3.0Hz [Earlier results used a “worst case” where the noise-generation mask was used rather than oscillator performance]
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 5)

Chain output Phy_Layer - TDEV



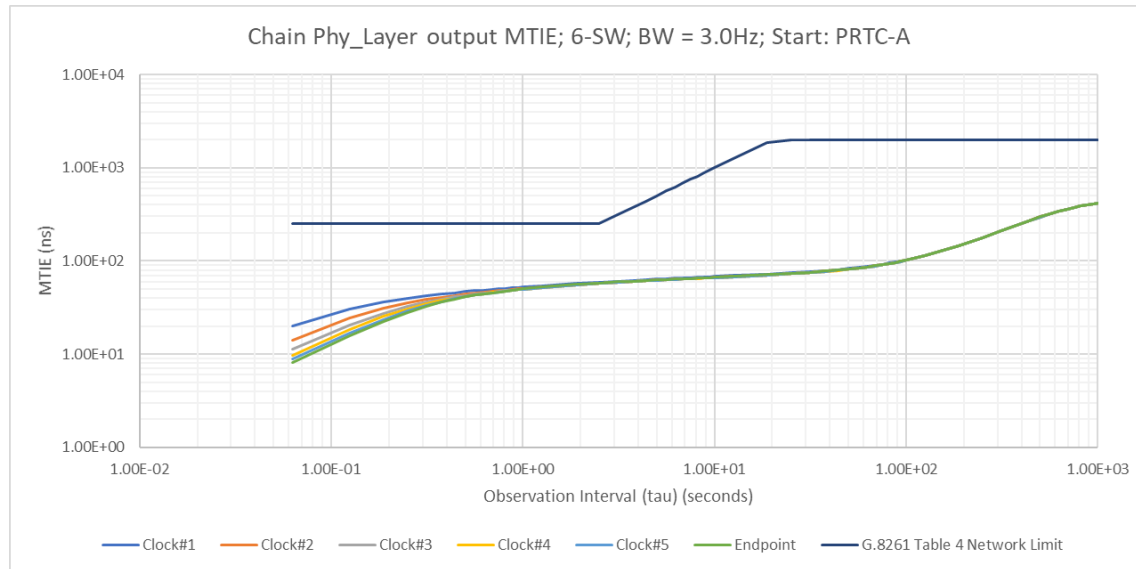
- ▶ Expected TDEV of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 0.05Hz
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 5)

Chain output Phy_Layer - TDEV



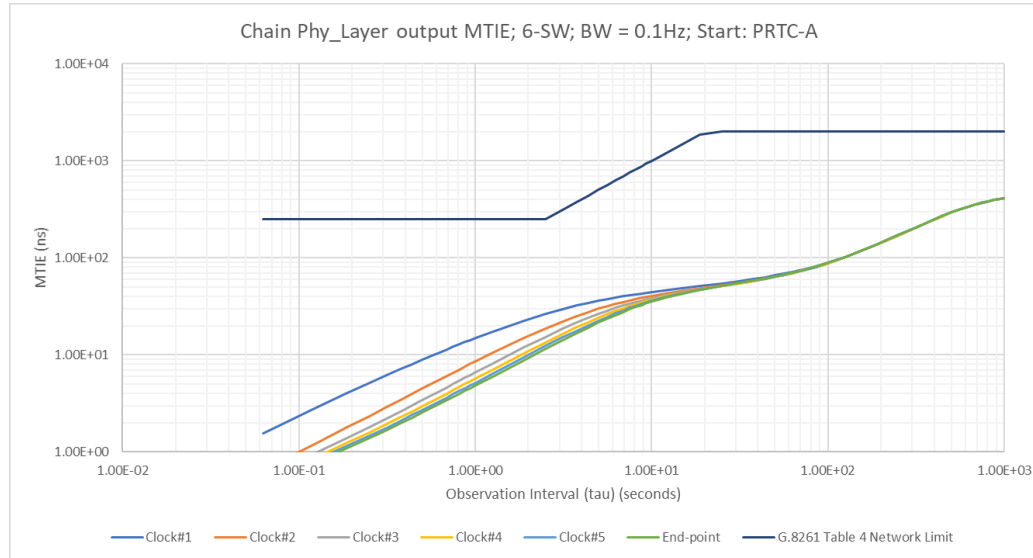
- Expected TDEV of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 0.1Hz
- Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- Mask is the network limit for Synchronous Ethernet (see G.8261 Table 5)

Chain output Phy_Layer - MTIE



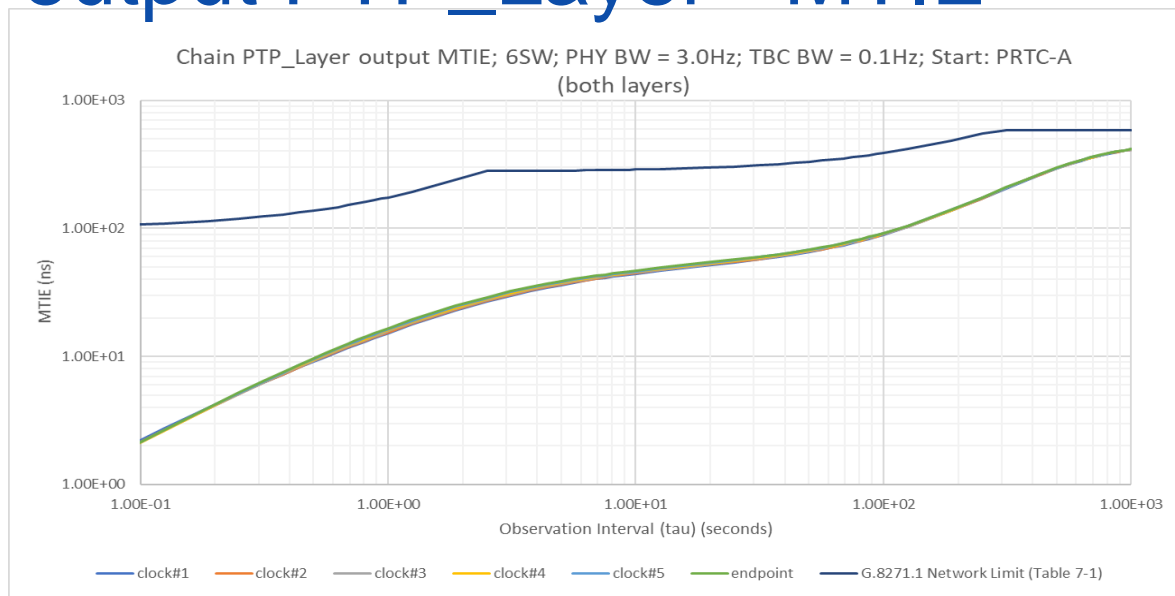
- ▶ Expected MTIE of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 3.0Hz [Earlier results used a “worst case” where the noise-generation mask was used rather than oscillator performance]
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 4)

Chain output Phy_Layer - MTIE



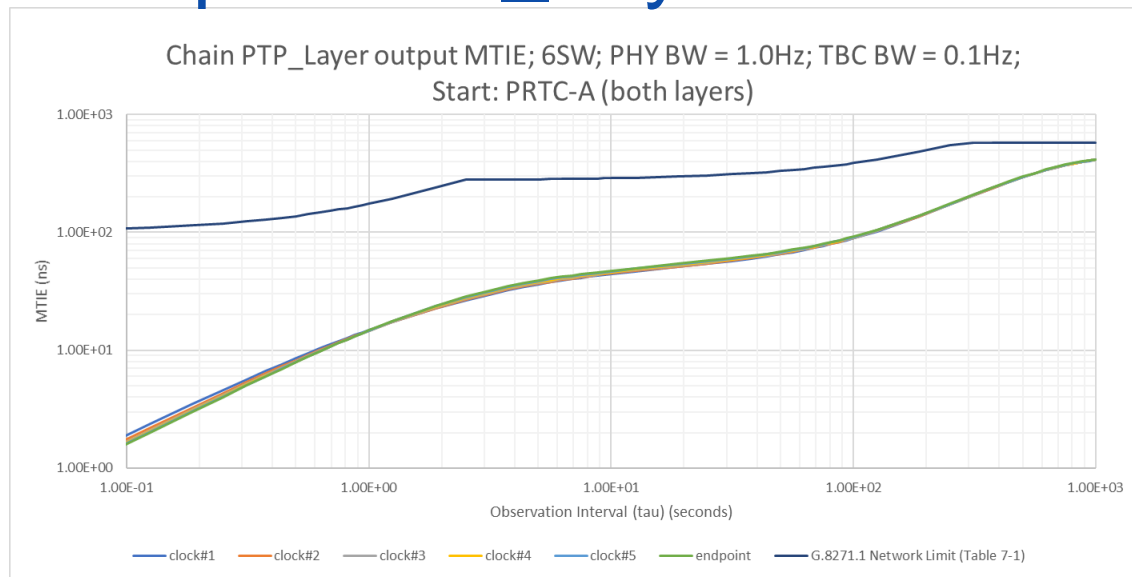
- ▶ Expected MTIE of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 0.1Hz
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 4)

Chain output PTP_Layer - MTIE



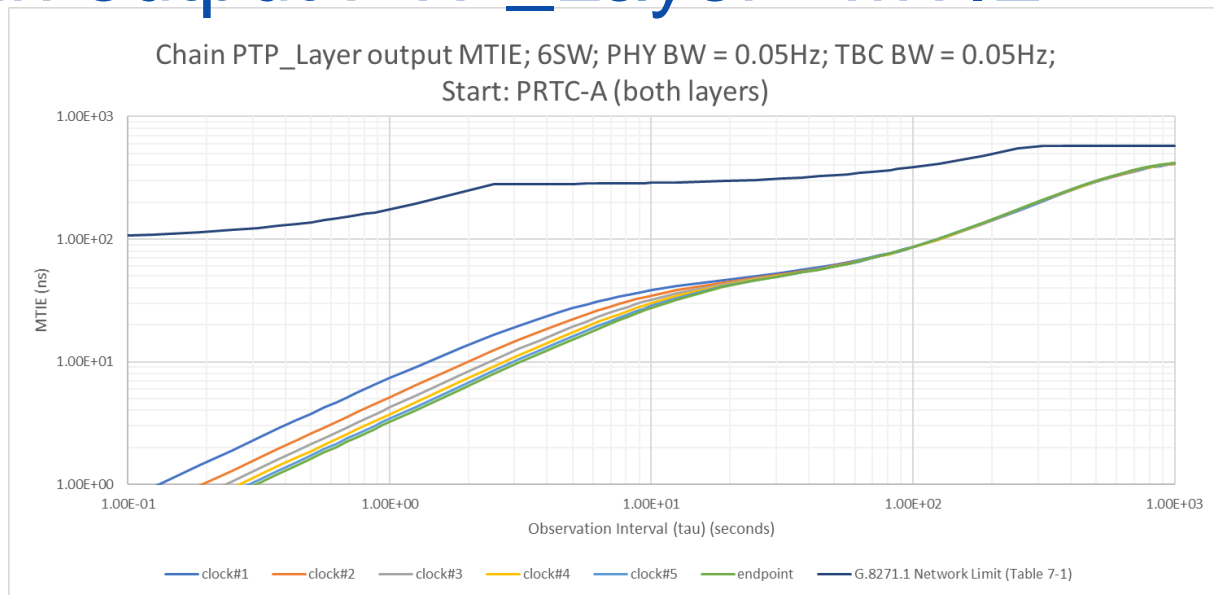
- ▶ Expected MTIE of clock chain (PTP_layer) (including endpoint) using oscillator with assumed performance ($TDEV(@1s) = 0.1ns$) for phy_layer bandwidth of 3.0Hz and PTP_layer bandwidth of 0.1Hz
- ▶ Start-of-chain is PRTC-A for both phy_layer and PTP_layer
- ▶ Mask is the network limit from G.8271.1 (Table 7-1)
- ▶ Time-stamp granularity of 8ns

Chain output PTP_Layer - MTIE



- ▶ Expected MTIE of clock chain (PTP_layer) (including endpoint) using oscillator with assumed performance ($TDEV(@1s) = 0.1ns$) for phy_layer bandwidth of 1.0Hz and PTP_layer bandwidth of 0.1Hz
- ▶ Start-of-chain is PRTC-A for both phy_layer and PTP_layer
- ▶ Mask is the network limit from G.8271.1 (Table 7-1)
- ▶ Time-stamp granularity of 8ns

Chain output PTP_Layer - MTIE



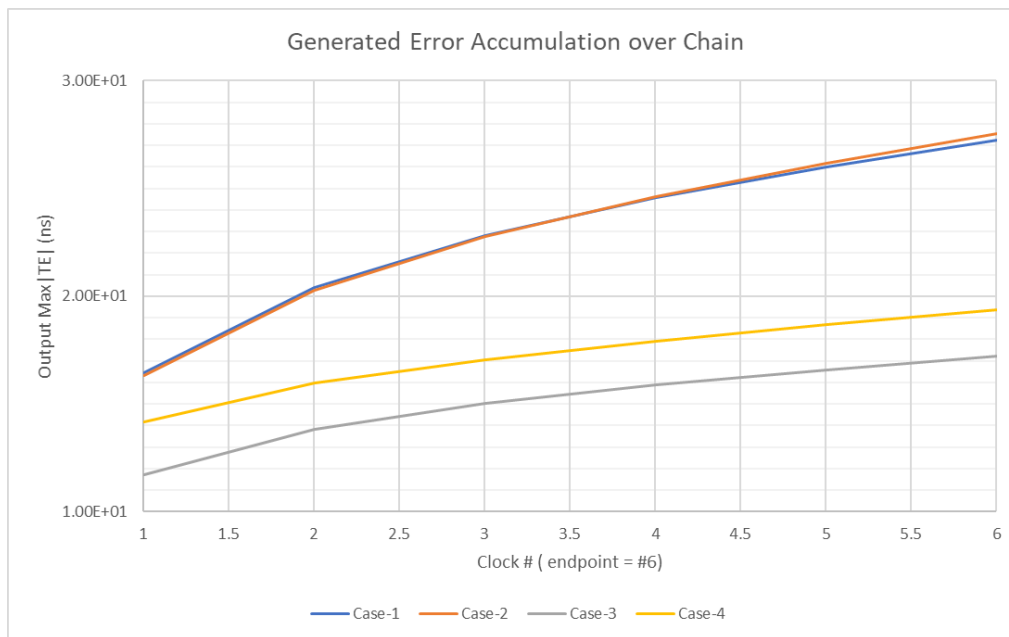
- ▶ Expected MTIE of clock chain (PTP_layer) (including endpoint) using oscillator with assumed performance ($TDEV(@1s) = 0.1ns$) for phy_layer bandwidth of 0.05Hz and PTP_layer bandwidth of 0.05Hz
- ▶ Start-of-chain is PRTC-A for both phy_layer and PTP_layer
- ▶ Mask is the network limit from G.8271.1 (Table 7-1)
- ▶ Time-stamp granularity of 8ns

Chain Time Error Accumulation

- ▶ To see accumulation of time error generated by the chain, the starting point (PRTC-A/T-GM) introduces only time-stamping error
- ▶ Physical Layer starts with PRTC-A (essentially G.811) at its limit (TDEV mask)
- ▶ Time error estimated with 0.1Hz measurement filter
- ▶ Time-stamp granularity assumed to be 8ns
- ▶ 4 cases considered with different choices of filter bandwidths

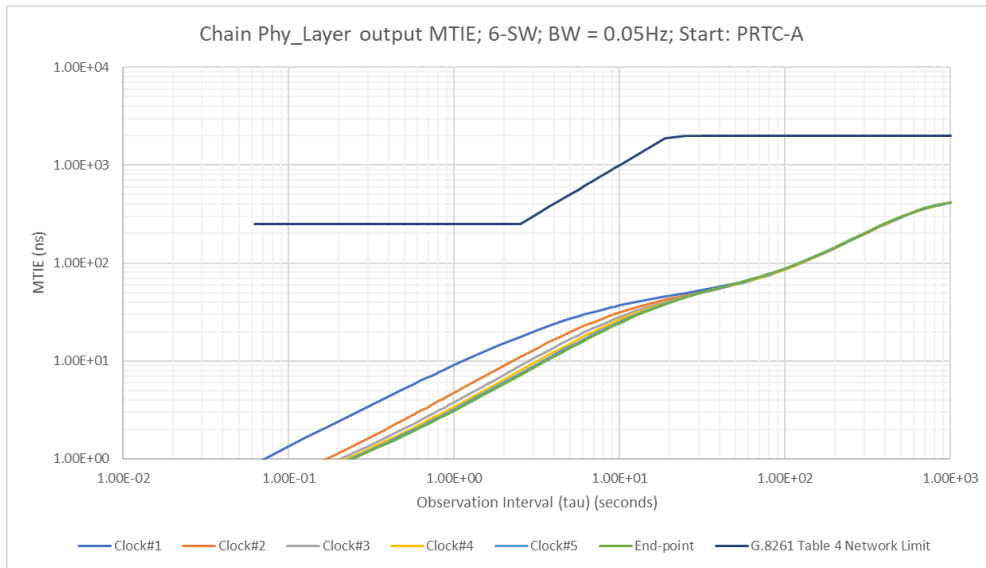
	Phy_layer bandwidth	PTP_layer bandwidth
Case #1	3.0 Hz	0.1 Hz
Case #2	1.0 Hz	0.1 Hz
Case #3	0.1 Hz	0.1 Hz
Case #4	0.05 Hz	0.05 Hz

Chain Time Error Accumulation



- ▶ Error accumulation shows an interaction between choices of bandwidth
 - Case #3 has 0.1Hz bandwidth for both layers
- ▶ The optimal choice of bandwidths will depend on assumed oscillator performance

Chain output Phy_Layer - MTIE



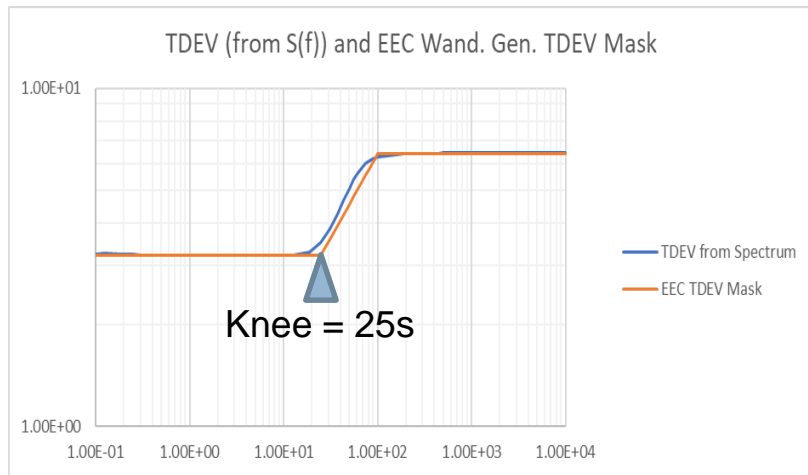
- ▶ Expected MTIE of clock chain (phy_layer) using oscillator with assumed performance for bandwidth of 0.1Hz
- ▶ Start-of-chain is PRTC-A (equivalent of G.811 for phy_layer)
- ▶ Mask is the network limit for Synchronous Ethernet (see G.8261 Table 4)

EEC/eEEC Wander Generation TDEV Mask

$$S_x(f) = \left(\frac{G}{f} \cdot (a_1 \cdot H_1(f) + a_2 \cdot H_2(f)) \right) \cdot C(f)$$

$H_1(f)$: 2nd order high-pass filter; $H_2(f)$: 3rd order low-pass filter; $(1/f)$: spectral density shape of flicker phase noise

For eEEC: power-spectral-density = *scale-factor**(EEC_PSD)



scale-factor = α

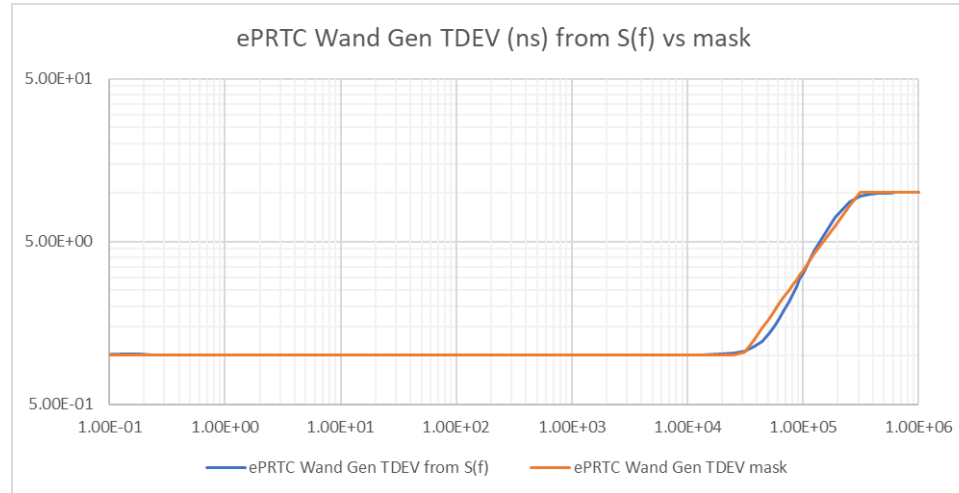
In G.8262.1 (eEEC) the appropriate factor is ~0.2

$$S_x(f) = \alpha \cdot \left(\frac{G}{f} \cdot (a_1 \cdot H_1(f) + a_2 \cdot H_2(f)) \right) \cdot C(f)$$

ePRTC Wander Generation TDEV Mask

$$S_x(f) = \left(\frac{G}{f}\right) \cdot (1 + 100.0 \cdot H_L(f)) \cdot C(f)$$

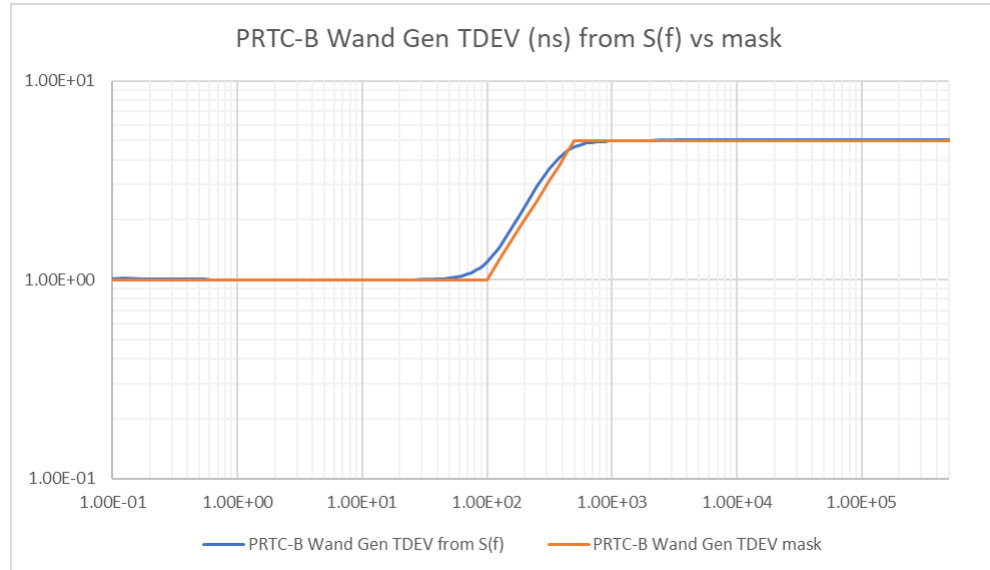
- ▶ $G \sim 0.44$.
- ▶ $H_L(f)$ is a fourth-order low-pass filter characteristic with 3dB cut-off frequency = 0.000002 Hz.
- ▶ $(1/f)$: spectral density shape of flicker phase noise



PRTC-B Wander Generation TDEV Mask

$$S_x(f) = \left(\frac{G}{f} \cdot (a_1 \cdot H_1(f) + a_2 \cdot H_2(f)) \right) \cdot C(f)$$

- $H_1(f)$ is a 2nd-order high-pass filter with 3-dB frequency of $\sim 0.001\text{Hz}$ and $a_1 = 1.0^2$;
- $H_2(f)$ is a 3rd-order low-pass filter with 3-dB frequency of $\sim 0.0012\text{Hz}$ and $a_2 = 5.05^2$;
- $G \sim 0.44$.





Thank You!

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